

**FEATURES**

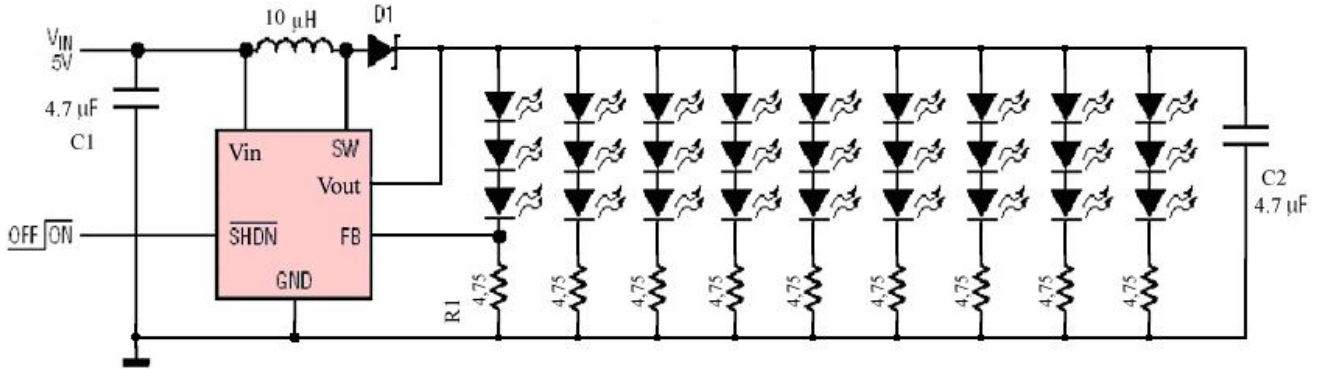
- Inherently Matched LED Current
- Drives Up to 27 LEDs from a 5V Supply
- 36V Rugged Bipolar Switch
- Fast 1.2MHz Switching Frequency
- VOUT(MAX)=30V

**DESCRIPTION**

The  $\mu$ PT111 is a step-up DC/DC converter specifically designed to drive white LEDs with a constant current. The device can drive up to 27 LEDs from a 5V supply.

Additional feature include output voltage limiting when LEDs are disconnected.

**Figure 1. TYPICAL APPLICATION**

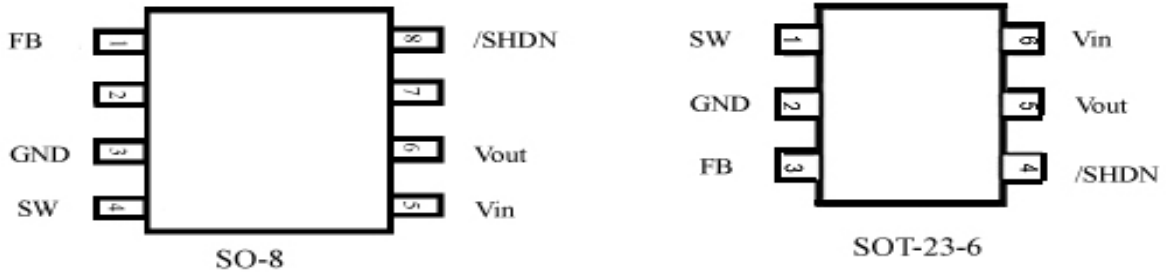


**ABSOLUTE MAXIMUM RATINGS (Note 1)**

Input Voltage (VIN)	10V	Operating Temperature Range	0°C to 70°C
SW Voltage	36V	Maximum Junction Temperature	125°C
FB Voltage	10V	Storage Temperature Range	-65°C to 150°C
/SHDN Voltage	10V	Lead Temperature (Soldering, 10 sec)	300°C

**Note 1:** Absolute Maximum Ratings are those values beyond which the life of the device may be impaired.

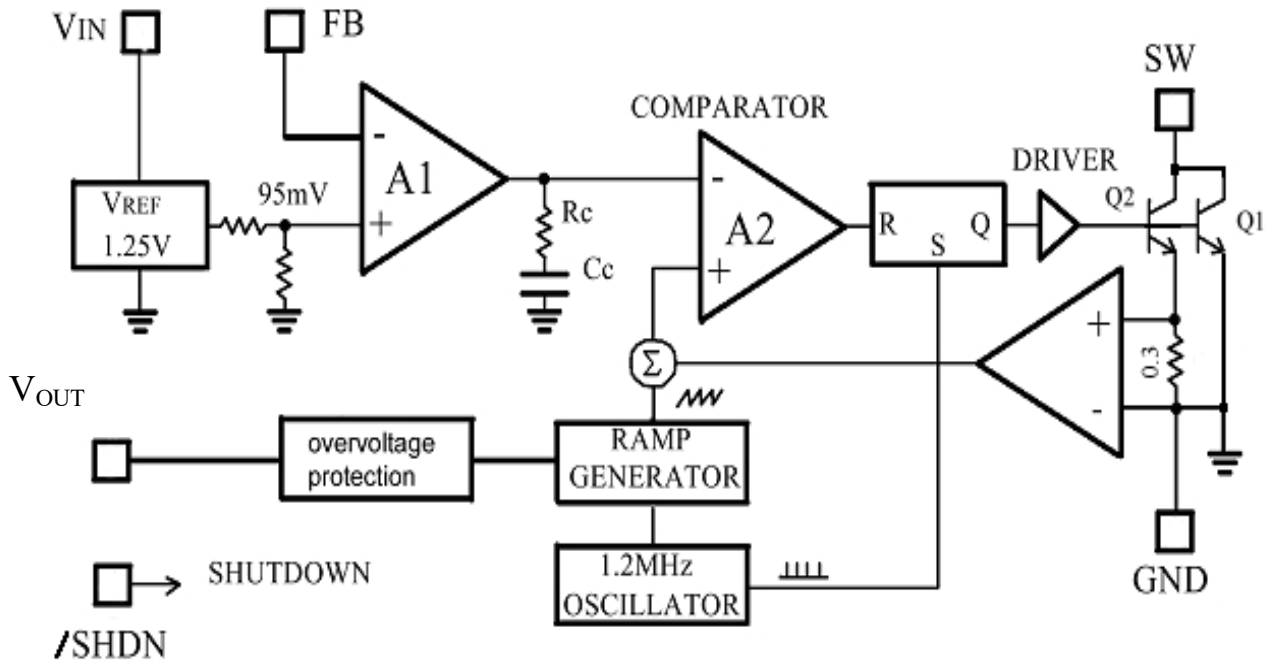
**PIN CONFIGURATION**



**ELECTRICAL CHARACTERISTICS** TA = 25°C, VIN = 5V, V /SHDN = 5V, unless otherwise noted.

PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Minimum Operating Voltage		2.5			V
Maximum Operating Voltage				10	V
Feedback Voltage	ILOAD = 180mA, VIN=5V	83	95	107	mV
	ILOAD = 100mA, VIN=5V	86	95	104	
FB Pin Bias Current		10	45	100	nA
Supply Current	/SHDN = 0V		2.1	3.0	mA
			0.1	1.0	
Switching Frequency		1.1	1.3	1.6	MHz
Maximum Duty Cycle		85	90		%
Switch Current Limit			650		mA
Switch VCESAT	ISW = 250mA		350		mV
Switch Leakage Current	VSW = 5V		0.01	5	µA
/SHDN Voltage High		1.5			V
/SHDN Voltage Low				0.4	V
/SHDN Pin Bias Current			65		µA
OVP Threshold			29		V

Figure 2. BLOCK DIAGRAM <A) \*\*\*



## OPERATION

The PT1111 uses a constant frequency, current mode control scheme to provide excellent line and load regulation. Operation can be best understood by referring to the block diagram in Figure 2. At the start of each oscillator cycle, the RS latch is set, which turns on the power switch Q1. A voltage proportional to the switch current is added to a stabilizing ramp and the resulting sum is fed into the positive terminal of the PWM comparator A2. When this voltage exceeds the level at the negative input of A2, the RS latch is reset turning off the power switch. The level at the negative input of A2 is set by the error amplifier A1, and is simply an amplified version of the difference between the feedback voltage and the reference voltage of 95mV. In this manner, the error amplifier sets the correct peak current level to keep the output in regulation. If the error amplifier's output increases, more current is delivered to the output; if it decreases, less current is delivered.

## APPLICATIONS INFORMATION

### Inductor Selection

A 10μH inductor is recommended for most PT1111 applications. Although small size and high efficiency are major concerns, the inductor should have low core losses at 1.2MHz and low DCR (copper wire resistance).

### Capacitor Selection

The small size of ceramic capacitors makes them ideal for PT1111 applications. X5R and X7R types are

recommended because they retain their capacitance over wider voltage and temperature ranges than other types such as Y5V or Z5U. A 4.7μF input capacitor and a 4.7μF output capacitor are sufficient for most PT1111 applications.

### Diode Selection

Schottky diodes, with their low forward voltage drop and fast reverse recovery, are the ideal choices for PT1111 applications. The forward voltage drop of a Schottky diode represents the conduction losses in the diode, while the diode capacitance ( $C_T$  or  $C_D$ ) represents the switching losses. For diode selection, both forward voltage drop and diode capacitance need to be considered. Schottky diodes with higher current ratings usually have lower forward voltage drop and larger diode capacitance, which can cause significant switching losses at the 1.2MHz switching frequency of the PT1111. A Schottky diode rated at 1000mA is sufficient for most PT1111 applications.

### LED Current Control

The LED current is controlled by the feedback resistor (R1 in Figure 1). The feedback reference is 95mV. The LED current is  $95\text{mV}/R1$ . The formula and table 3 for R1 selection are shown below.  $R1 = 95\text{mV}/I_{\text{LED}}$

Table 1. R1 Resistor Value Selection

$I_{\text{LED}} \text{ (mA)}$	$R1 \text{ (}\Omega\text{)}$
5	19.1
10	9.53
12	7.87
15	6.34
20	4.75